

BUH515D

HIGH VOLTAGE FAST-SWITCHING NPN POWER TRANSISTOR

- STMicroelectronics PREFERRED SALESTYPE
- HIGH VOLTAGE CAPABILITY
- FULLY INSULATED PACKAGE (U.L. COMPLIANT) FOR EASY MOUNTING
- NPN TRANSISTOR WITH INTEGRATED FREEWHEELING DIODE

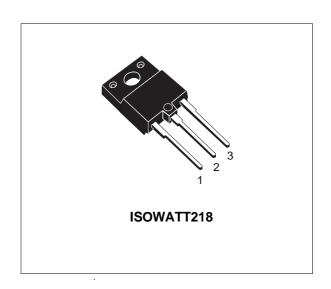
APPLICATIONS:

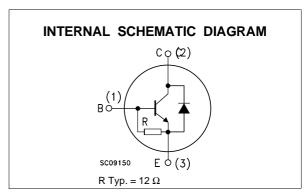
 HORIZONTAL DEFLECTION FOR COLOUR TVS

DESCRIPTION

The BUH515D is manufactured using Multiepitaxial Mesa technology for cost-effective high performance and uses a Hollow Emitter structure to enhance switching speeds.

The BUH series is designed for use in horizontal deflection circuits in televisions and monitors.





ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _{CBO}	Collector-Base Voltage (I _E = 0)	1500	V
V _{CEO}	Collector-Emitter Voltage (I _B = 0)	700	V
V _{EBO}	Emitter-Base Voltage (I _C = 0)	5	V
Ic	Collector Current	8	А
I _{CM}	Collector Peak Current (t _p < 5 ms)	15	А
lΒ	Base Current	5	А
I _{BM}	Base Peak Current (t _p < 5 ms)	8	A
P _{tot}	Total Dissipation at T _c = 25 °C	50	W
V _{isol}	Insulation Withstand Voltage (RMS) from All	2500	V
	Three Leads to Exernal Heatsink		
T _{stg}	Storage Temperature	-65 to 150	°C

July 2002 1/7

BUH515D

THERMAL DATA

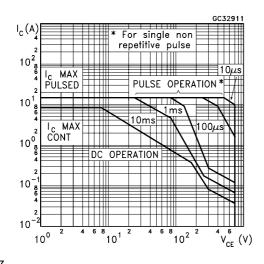
R _{thj-case} Thermal Resistance Junction-case	Max	2.5 °C/W	Junction-case	Resistance	Thermal	R _{thj-case}
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ELECTRICAL CHARACTERISTICS (T_{case} = 25 °C unless otherwise specified)

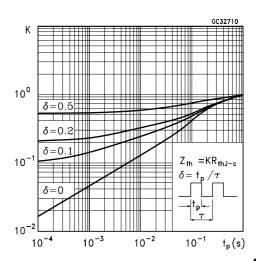
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I _{CES}	Collector Cut-off Current (V _{BE} = 0)	V _{CE} = 1300 V V _{CE} = 1500 V V _{CE} = 1500 V T _j = 125 °C			10 0.2 2	μA mA mA
I _{EBO}	Emitter Cut-off Current (I _C = 0)	V _{EB} = 5 V			200	mA
$V_{CE(sat)^*}$	Collector-Emitter Saturation Voltage	I _C = 5 A I _B = 1.25 A			1.5	V
$V_{BE(sat)^*}$	Base-Emitter Saturation Voltage	I _C = 5 A I _B = 1.25 A			1.3	V
h _{FE} *	DC Current Gain	$I_{C} = 5 \text{ A}$ $V_{CE} = 5 \text{ V}$ $I_{C} = 5 \text{ A}$ $V_{CE} = 5 \text{ V}$ $T_{j} = 100 ^{\circ}\text{C}$	5 3		10	
t _s	RESISTIVE LOAD Storage Time Fall Time	V _{CC} = 400 V I _C = 5 A I _{B1} = 1.5 A I _{B2} = -2.5 A		2.4 170	3.6 260	μs ns
t _s t _f	INDUCTIVE LOAD Storage Time Fall Time	$\begin{aligned} &\text{Ic} = 5 \text{ A} & \text{f} = 15625 \text{ Hz} \\ &\text{I}_{B1} = 1.25 \text{ A} & \text{I}_{B2} = -2.5 \text{ A} \\ &\text{V}_{\text{ceflyback}} = 1050 \sin\!\left(\!\frac{\pi}{10}10^6\!\right)\!t & \text{V} \end{aligned}$		3.5 450		μs ns
VF	Diode Forward Voltage	I _F = 5 A			2	V

^{*} Pulsed: Pulse duration = 300 μs, duty cycle 1.5 %

Safe Operating Area

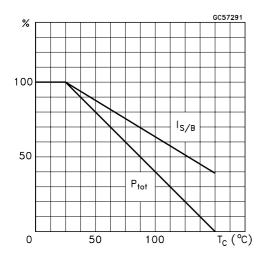


Thermal Impedance

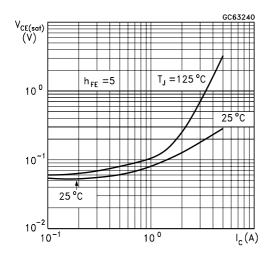


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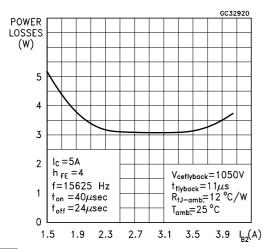
Derating Curve



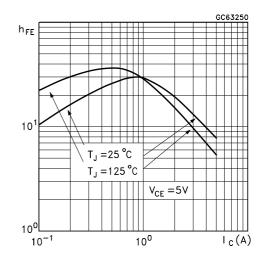
Collector Emitter Saturation Voltage



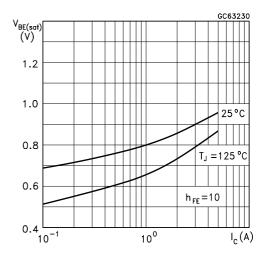
Power Losses at 16 KHz



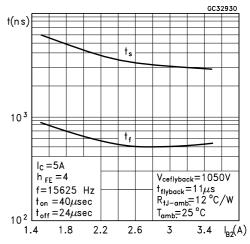
DC Current Gain



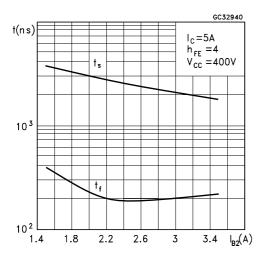
Base Emitter Saturation Voltage



Switching Time Inductive Load at 16KHz (see figure 2)



Switching Time Resistive Load



BASE DRIVE INFORMATION

In order to saturate the power switch and reduce conduction losses, adequate direct base current l_{B1} has to be provided for the lowest gain h_{FE} at $100\,^{\circ}C$ (line scan phase). On the other hand, negative base current l_{B2} must be provided to turn off the power transistor (retrace phase). Most of the dissipation, especially in the deflection application, occurs at switch-off. Therefore it is essential to determine the value of l_{B2} which minimizes power losses, fall time t_f and, consequently, $T_j.$ A new set of curves have been defined to give total power losses, t_s and t_f as a function of l_{B2} at 16 KHz frequencies for choosing the optimum negative drive. The test circuit is illustrated in fig. 1.

Inductance L_1 serves to control the slope of the negative base current I_{B2} to recombine the excess carrier in the collector when base current is still present, this avoid any tailing phenomenon in the collector current.

The values of L and C are calculated from the following equations:

$$\frac{1}{2}L(I_C)^2 = \frac{1}{2}C(V_{CEfly})^2$$
$$\omega = 2\pi f = \frac{1}{\sqrt{LC}}$$

Where I_{C} = operating collector current, V_{CEfly} = flyback voltage, f= frequency of oscillation during retrace.

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Figure 1: Inductive Load Switching Test Circuit

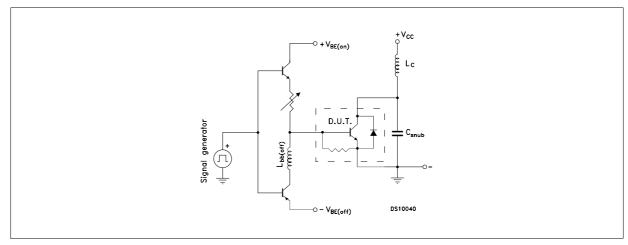
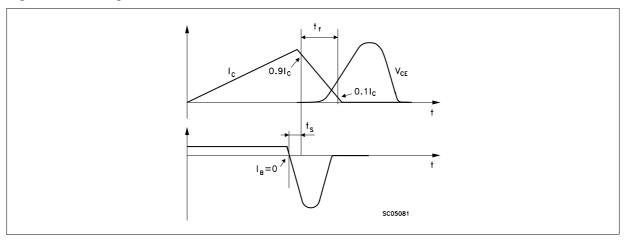


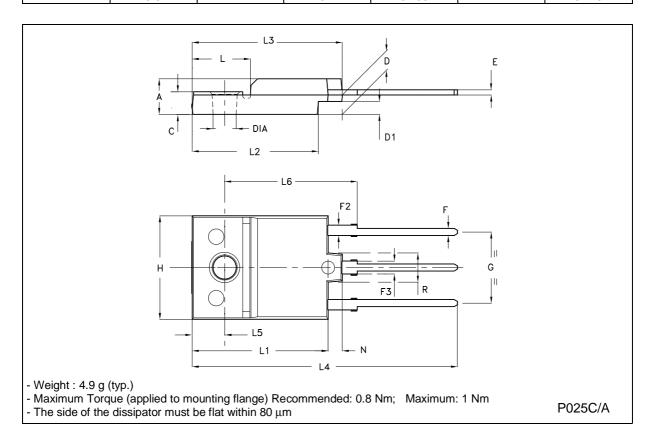
Figure 2: Switching Waveforms in a Deflection Circuit



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ISOWATT218 MECHANICAL DATA

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α	5.35		5.65	0.211		0.222
С	3.30		3.80	0.130		0.150
D	2.90		3.10	0.114		0.122
D1	1.88		2.08	0.074		0.082
Е	0.75		0.95	0.030		0.037
F	1.05		1.25	0.041		0.049
F2	1.50		1.70	0.059		0.067
F3	1.90		2.10	0.075		0.083
G	10.80		11.20	0.425		0.441
Н	15.80		16.20	0.622		0.638
L		9			0.354	
L1	20.80		21.20	0.819		0.835
L2	19.10		19.90	0.752		0.783
L3	22.80		23.60	0.898		0.929
L4	40.50		42.50	1.594		1.673
L5	4.85		5.25	0.191		0.207
L6	20.25		20.75	0.797		0.817
N	2.1		2.3	0.083		0.091
R		4.6			0.181	
DIA	3.5		3.7	0.138		0.146



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